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## Article

# Power Management of a Hybrid Micro-Grid with Photovoltaic Production and Hydrogen Storage

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**Abstract:** To deal with energy transition due to climate change and a rise in average global temperature, photovoltaic (PV) conversion appears to be a promising technology in sunny regions. However, PV production is directly linked with weather conditions and the day/night cycle, which makes it intermittent and random. Therefore, it makes sense to combine it with Energy Storage Systems (ESS) to ensure long-term energy availability for non-interconnected micro-grids. Among all technological solutions, electrolytic hydrogen produced by renewable energies seems an interesting candidate. In this context, this paper proposes a control strategy dedicated to hydrogen storage integration in micro-grids for a better use of PV production. The objective is to optimize the management of the micro-grid with proton exchange membrane Fuel Cell (FC), alkaline Electrolyzer (El), lithium-ion Batteries Energy Storage System (BESS) and PV, according to the system state and PV production intermittency. First, a control strategy based on a Distributed explicit Model Predictive Control (DeMPC) is developed to define current references for FCs, Els and batteries. Secondly, the performance of the control strategy is validated in simulation and confirmed on a Power-Hardware-in-the-Loop test bench.

**Keywords:** hybridization; distributed explicit model predictive control; fuel cell; electrolyzer; hydrogen; power management system; power-hardware-in-the-loop



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## 1. Introduction

In the past few years, the various actors in the energy field (companies, researchers and public authorities) have highlighted a common interest in electricity generation produced with green energy. A lot of research in the field of renewable energies has made it possible to integrate them into the electrical grid. However, the energy produced by photovoltaic panels and wind turbines depend on climatic conditions, thus leading to high intermittency and random production. As a result, the generation–consumption balance cannot always be ensured, which can impact the electrical grid stability, especially in non-interconnected zones. Therefore, one solution consists of combining them with long-term Energy Storage System (ESS) to deal with the excess or deficit energy [1–3]. The most widely used storage systems is batteries. However, these systems have specific drawbacks depending on the technologies [4]. For instance, lead-acid batteries have a high weight and a shorter lifetime than other technologies [5]. Lithium batteries are very expensive and require specific conditions for transport and use due to their chemical composition [5,6]. Among all technological solutions, electrolytic hydrogen produced by renewable energies seems an interesting candidate for a deferred use with Fuel Cell (FC) systems. It has the advantage of a hydrogen production without carbon dioxide emission as well as the decoupling of storage capacity and power. Indeed, battery capacity storage and power depend directly on

the number of cells used. In contrast, hydrogen storage capacity depends on the considered technology (e.g., solid, liquid or gaseous), while power generation depends on FC itself. Combining hydrogen and batteries storage can take advantage of both technology: high power density for the batteries and high energy density from the hydrogen system [7]. Nevertheless, FC and Electrolyzer (El) can suffer from severe degradation if their operating conditions are not respected, e.g., fast variations of the current, high and low operating powers, and so on [8–11]. In this context, ensuring reliable and accurate power response of all the components while respecting proper operating conditions remain a challenge.

Hydrogen storage systems are widely studied for transport applications. In Hilairat et al. [12], authors use the passivity methodology to control a FC and super-capacitor (SC) hybrid system. The controller computes the reference currents for each unit to regulate the Direct Current (DC) bus voltage and ensure good operating conditions of the FC. Simulation and experimental results show good performance. A Mixed Droop Control Strategy to allocate power to the FC and the SCs considering current dynamic constraints is developed in Chen et al. [13]. The control strategy is validated in simulation and in a Hardware-in-the-Loop real-time simulation platform. An Extremum Seeking method is proposed in Zhou et al. [14] to control an FC and battery hybrid system for electrical vehicles. The effectiveness of the control is validated with experimental results.

Hydrogen is also under consideration for micro-grids. In Sun et al. [15], authors propose two basic coordinated control strategies for a hybrid energy system composed of batteries, wind turbines, FCs and photovoltaic panels (PVs) connected to the grid. The performance of the proposed control is assessed in simulation. The authors in Sharma et al. [16] propose a conditional statements power management algorithm with three operation modes in function of the difference between PV and load powers. The hybrid system is composed of FC, batteries and PV. The algorithm computes the reference currents for each unit of the micro-grid while taking into account constraints on current dynamics of each element. The authors demonstrate the effectiveness of the strategy using simulation and experimentation results. In the same context, in Mane et al. [17] the authors propose a flowchart with conditional statement functions of PV, load, and FC powers, as well as the battery state of charge. Simulation results show the effectiveness of the power management strategy. A multiple-modes power distribution strategy is used in Li et al. [18] for a DC micro-grid composed of batteries, FC, wind turbines, and PV, where the FC power increases in steps of 100 W depending on renewable source availability and load power demand. The strategy is experimentally validated. In Mane et al. [19], the authors proposed a non-linear controller based on Lure–Lyapunov formulation for a hybrid system composed of FC and SCs. Experimental results are provided. Even if several studies have been conducted regarding the integration of hydrogen vector in different systems, few studies take explicitly into account all the component constraints of the micro-grid in the controller.

The main drawback of these approaches is that they do not allow explicit integration of the minimum and maximum current constraints as well as the maximum current dynamics of the FC and the electrolyzer into the controller without deteriorating the stability of the closed-loop system. In this context, in Mariethoz et al. [20] the authors use a Model Predictive Control (MPC) that explicitly incorporates constraints into the controller design. MPC has the advantage of being able to easily manage multi-variable systems and their constraints. In recent years, the existence of mathematical models of power electronics has made it possible to use MPC in electrical systems. However, the computational load makes the implementation difficult with a low time-step. To overcome this issue, the optimization problem can be solved offline by multi-parametric programming to reduce the implementation to a few computations and a look-up table. This approach reduces the computational cost and allows the MPC to be used with a low time-step [21].

This study concerns the integration of hydrogen produced by decarbonated energy from PV panels for a better use of renewable energies in an islanded micro-grid to supply public buildings [22], such as the SAGES project in Mafate, La Réunion. The main issues preventing the use of hydrogen as an energy vector in micro-grids is the low reliability

and high cost of this technology due to the brittleness and complex usage. However, an adapted controller allows the respecting of the constraints on the system and to improve the remaining useful life of the components. In this context, a Distributed explicit Model Predictive Control (DeMPC) has been applied to control a complex hydrogen-energy system composed of a FC/El associated with long-term hydrogen storage, batteries for short-term storage and PV panels as main power source. The Power Management System (PMS) explicitly takes into account each unit operating constraints (maximum/minimum current, response time, etc.) [20]. Moreover, it is possible to take into account the architecture of the system in real time when the power electronic topology is reconfigured in case of failure [23] or units (FC or El) commitment in case of power partitioning, as well as to change the tuning parameters according to the FC/El state of health based on diagnosis/prognosis tools [24]. The performance of the PMS is assessed experimentally on a Power-Hardware-In-the-Loop (PHIL) test bench to confirm the effectiveness of the proposition. The main purpose of this work is the integration of the PMS with the DeMPC controller to define the reference currents of the elements in a complex system. The objective of the DeMPC is the control of the DC bus voltage and the State of Charge (SoC) of the battery to ensure the stability of the micro-grid without damaging the FC and the electrolyzer. The principal contributions of this work are:

- The implementation of an DeMPC-based control strategy that explicitly takes into account the constraints of the elements in complex systems such as a micro-grid with a hydrogen system,
- The performance of tests of the control strategy on a PHIL test bench to show the applicability of the strategy on real systems.

This paper is organized as follows. The PMS design and the simulation results are given in Section 2. Experimental results are given in Section 3 and the conclusion is given in Section 5.

## 2. Power Management System Design and Simulation

### 2.1. Micro-Grid Modeling

The micro-grid studied in this paper is composed of Batteries Energy Storage System (BESS), PV, a FC stack and an El. Each element is connected to a DC/DC converter as shown in Figure 1.

#### 2.1.1. Photovoltaic Model

The PV power depends on the solar radiation and PV cell temperature. A one-diode equivalent scheme [25,26] is identified based on TENESOL (Lyon, France) TE1700 manufacturer data [27]. The PV module has its own Maximum Power Point Tracking (MPPT) controller independent of the PMS.

#### 2.1.2. Battery Model

A 3.7 V and 800 mAh Li-ion battery cell is modeled by a second-order Thevenin equivalent scheme [28]. The model takes into account experimental data to extract model parameters that depend on the battery SoC, computed by the regular Coulomb model according to the battery current Equation (1).

$$SoC(t) = SoC(t = 0) + \frac{1}{C_{bat}} \int_0^t i_{bat}(\tau) d\tau \quad (1)$$

#### 2.1.3. Fuel Cell and Electrolyzer Models

The FC voltage  $v_{fc}$  is computed based on a 5-order polynomial equation of the FC current obtained from experimental tests on a 1.2 kW, 46 A and 26 V Nexa Ballard FC [12]. In the same way, the El voltage  $v_{el}$  depends on the El current  $i_{el}$  based on [29]. The  $v_{el}$  equation parameters have been fitted based on experimental results of an H2 IGen

300/1/25 manufactured by Vandenberg (currently, Hydrogenics, Mississauga, Canada) 5 kW, 43 V and 120 A Alkaline Electrolyzer.

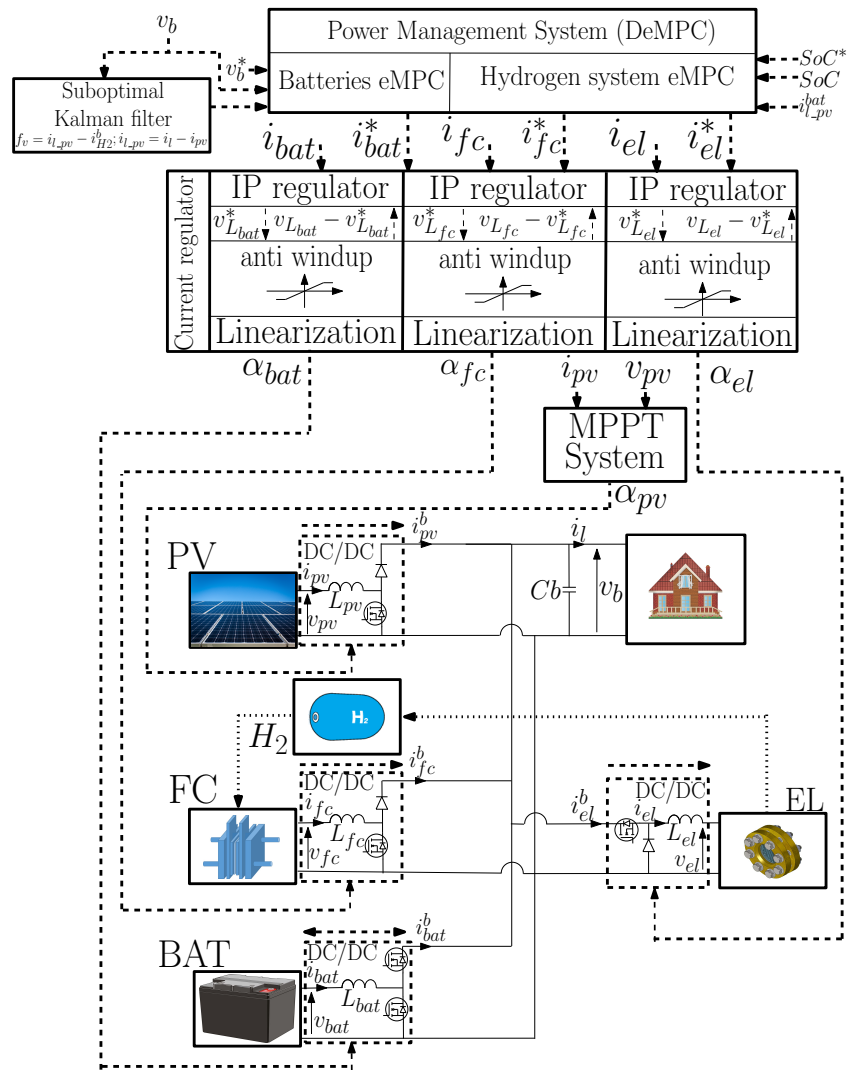


Figure 1. Micro-grid structure scheme.

#### 2.1.4. Overall System Model

Two types of DC/DC converters are used in this study—a buck converter for the EL and boost converters for all other units. The behavior of the system in Figure 1 can be described by a full order non-linear system Equations (2)–(7) where all the DC/DC converters are represented by an average model.

$$\frac{dv_b(t)}{dt} = \frac{1}{C_b} [(1 - \alpha_{fc}(t))i_{fc}(t) + (1 - \alpha_{bat}(t))i_{bat}(t) + (1 - \alpha_{pv}(t))i_{pv}(t) - \alpha_{el}(t)i_{el}(t) - i_l(t)] \quad (2)$$

$$\frac{di_{fc}(t)}{dt} = \frac{-((1 - \alpha_{fc}(t))i_{fc}(t))v_b(t) + v_{fc}(t)}{L_{fc}} \quad (3)$$

$$\frac{di_{bat}(t)}{dt} = \frac{-((1 - \alpha_{bat}(t))i_{bat}(t))v_b(t) + v_{bat}(t)}{L_{bat}} \quad (4)$$

$$\frac{di_{pv}(t)}{dt} = \frac{-((1 - \alpha_{pv}(t))i_{pv}(t))v_b(t) + v_{pv}(t)}{L_{pv}} \quad (5)$$

$$\frac{di_{el}(t)}{dt} = \frac{-((1 - \alpha_{el}(t))i_{el}(t))v_b(t) + v_{el}(t)}{L_{el}} \quad (6)$$

$$\frac{dSoC(t)}{dt} = \frac{1}{C_{bat}} i_{bat}(t) \quad (7)$$

## 2.2. Power Management System Design

The PMS is based on a DeMPC that explicitly takes into account each component specific operating conditions and constraints.

First, the PVs has its proper Maximum Power Point Tracking (MPPT) command [30]. Secondly, the PMS is composed of an outer loop composed of two explicit Model Predictive Controls (eMPC), one for the BESS and another for the hydrogen system composed of the FC, El and hydrogen storage. The inner current control loops with Integral-Proportional (IP) controllers are used and tune so that the inner loops are fast compared to the dynamic of the outer loops based on the DeMPC. Therefore, based on the singular perturbation theorem [31], the system equations for the design of the outer loops (DeMPC) can be simplified as follows [32]:

$$\frac{dv_b(t)}{dt} = \frac{1}{C_b} \left[ \frac{v_{fc}(t)}{v_b(t)} i_{fc}(t) + \frac{v_{bat}(t)}{v_b(t)} i_{bat}(t) + \frac{v_{pv}(t)}{v_b(t)} i_{pv}(t) - \frac{v_{el}(t)}{v_b(t)} i_{el}(t) - i_l(t) \right] \quad (8)$$

$$\frac{dSoC(t)}{dt} = \frac{1}{C_{bat}} \left[ \frac{v_{fc}(t)}{v_{bat}(t)} i_{fc}(t) + \frac{v_{pv}(t)}{v_{bat}(t)} i_{pv}(t) - \frac{v_{el}(t)}{v_{bat}(t)} i_{el}(t) - \frac{v_b(t)}{v_{bat}(t)} i_l(t) \right] \quad (9)$$

Here, all currents are supposed to be properly regulated by the inner current control loops. It follows that  $i_{fc}(t)$ ,  $i_{bat}(t)$  and  $i_{el}(t)$  are supposed perfectly regulated and are the inputs of the reduced order system (8) and (9).

### 2.2.1. BESS eMPC

The BESS eMPC computes the BESS current reference that controls the DC bus voltage. In the system, batteries consume the excess energy and restore it in case of an energy deficit when the other units cannot react. The PV and load currents ( $i_{pv}$  and  $i_l$ ) are considered to be perturbations in the eMPC and the difference between these two currents is called  $i_{l\_pv}$  for simplification.

- System equations

The battery eMPC is based on the linearized and discretized Equation (8) as follows.

$$v_{b,k+1} = v_{b,k} + \frac{T_{s_{bat}}}{C_b} \left[ i_{bat,k}^b + i_{H2,k}^b - i_{l_{pv},k} \right] \quad (10)$$

with:

$$i_{l_{pv},k} = i_l - i_{pv,k}^b \quad (11)$$

$$i_{H2,k}^b = i_{fc,k}^b + i_{el,k}^b \quad (12)$$

$$i_{pv,k}^b = \frac{v_{pv,k}}{v_{b,k}} i_{pv,k} \quad (13)$$

$$i_{fc,k}^b = \frac{v_{fc,k}}{v_{b,k}} i_{fc,k} \quad (14)$$

$$i_{bat,k}^b = \frac{v_{bat,k}}{v_{b,k}} i_{bat,k} \quad (15)$$

$$i_{el,k}^b = \frac{v_{el,k}}{v_{b,k}} i_{el,k} \quad (16)$$

where  $v_b$  is the state variable,  $i_{bat,k}^b$  is the control input and  $i_{H2}^b - i_{l_{pv}}$  is considered to be a perturbation.  $T_{s_{bat}}$  is the eMPC sampling time fixed at 500  $\mu$ s. At the end, the battery reference current value is sent to the battery current controller based on the inversion of Equation (15):

$$i_{bat,k}^* = \frac{v_{b,k}}{v_{bat,k}} i_{bat,k}^b \quad (17)$$

- System constraints

The only constraints taken into account for batteries are the minimum and maximum currents expressed in the DC bus point of view as follows:

$$i_{bat,min,k}^b \leq i_{bat,k}^b \leq i_{bat,max,k}^b \quad (18)$$

$$i_{bat,min,k}^b = \frac{v_{bat,k}}{v_{b,k}} i_{bat,min} \quad (19)$$

$$i_{bat,max,k}^b = \frac{v_{bat,k}}{v_{b,k}} i_{bat,max} \quad (20)$$

- Objective function

The objective function is written in the quadratic form. Its main role is to minimize the error between the DC bus voltage reference  $v_b^*$  and its measured value  $v_{b,k}$  at each time-step over the prediction horizon (the first term of the objective function in Equation (21)). Also, it minimizes the battery current seeing by the bus  $i_{bat,k}^b$  to stabilize it (second term of the objective function in Equation (21)). The objective function is written as:

$$J(v_{b,k}) = \min_{i_{bat}^b} \sum_{k=0}^{N_{bat}} \left[ (v_{b,k} - v_b^*) Q_{r_{bat}} (v_{b,k} - v_b^*)^T + \left( i_{bat,k}^b \right) \left( i_{bat,k}^b \right)^T \right] \quad (21)$$

$N_{bat}$  is the prediction horizon,  $Q_{r_{bat}}$  is the weight of the first objective function criterion and its value is fixed to respect the reference tracking while ensuring the battery current

stability. If  $Q_{r_{bat}}$  is too high, then there may be significant variations of  $i_{bat}$  between  $i_{bat,min}$  and  $i_{bat,max}$ . Its value is set to get  $v_b$  as close as possible to the reference value  $v_b^*$  while ensuring a minimum fluctuation in  $i_{bat}$ .

### 2.2.2. Hydrogen System eMPC

The ( $H_2$ ) system is composed of the FC and the El that does not work at the same time. Its current is considered to be the sum of the FC current and the El current Equation (23) computed by the eMPC to maintain the battery SoC at a define value. The El current is negative, and the FC current is positive.

- System equations

The hydrogen system eMPC is based on the linearization and discretization of Equation (9) where  $i_{H2,k}^{bat}$  is the input control. As with the BESS eMPC, the PV and load currents are merged in one variable noted  $i_{l-pv,k}^{bat}$ .

$$SoC_{bat,k+1} = SoC_{bat,k} + \frac{T_{sH2}}{C_{bat}(Ah)} \left[ i_{H2,k}^{bat} - i_{l-pv,k}^{bat}(k) \right] \quad (22)$$

with:

$$i_{H2,k}^{bat} = i_{fc,k}^{bat} + i_{el,k}^{bat} \quad (23)$$

$$i_{l-pv,k}^{bat} = i_{l,k}^{bat} - i_{pv,k}^{bat} \quad (24)$$

$$i_{pv,k}^{bat} = \frac{v_{pv,k}}{v_{bat,k}} i_{pv,k} \quad (25)$$

$$i_{fc,k}^{bat} = \frac{v_{fc,k}}{v_{bat,k}} i_{fc,k} \quad (26)$$

$$i_{l,k}^{bat} = \frac{v_{b,k}}{v_{bat,k}} i_{l,k} \quad (27)$$

$$i_{el,k}^{bat} = \frac{v_{el,k}}{v_{bat,k}} i_{el,k} \quad (28)$$

where  $SoC_{bat,k}$  is the state variable,  $i_{H2}^{bat}$  is the control input variable and  $i_b^{bat}$  is considered to be a perturbation.  $T_{sH2}$  is the eMPC sampling time fixed at 2 ms. FC and El current reference values are sent to the FC and El current controllers based on the inversion of Equations (26) and (28) respectively:

$$i_{fc,k}^* = \frac{v_{bat,k}}{v_{fc,k}} i_{fc,k}^{bat} \quad (29)$$

$$i_{el,k}^* = \frac{v_{bat,k}}{v_{el,k}} i_{el,k}^{bat} \quad (30)$$

- System constraints

Due to the design and physical phenomena into FC and El, specific operating conditions (response time, maximum power, etc.) need to be respected to avoid premature ageing. The eMPC needs to take into account the maximum and minimum currents (31) and the maximum and minimum current ramp rates (32). These parameters need to be written linearly to be integrated in the eMPC Equations (33)–(35):

$$i_{H2,min,k}^{bat} \leq i_{H2,k}^{bat} \leq i_{H2,max,k}^{bat} \quad (31)$$

$$\Delta i_{H2,max}^{bat} \leq \Delta i_{H2,k}^{bat} \leq \Delta i_{H2,max,k}^{bat} \quad (32)$$

$$i_{H2,min,k}^{bat} = \frac{v_{H2,k}}{v_{bat,k}} i_{H2,min} \quad (33)$$



$$i_{H2,max,k}^{bat} = \frac{v_{H2,k}}{v_{bat,k}} i_{H2,max} \quad (34)$$

$$\Delta i_{H2,max,k}^{bat} = \frac{v_{H2,k}}{v_{bat,k}} \Delta i_{H2,max} \quad (35)$$

- Objective function

Like the BESS eMPC, the objective function is written in the quadratic form. The main criterion is to ensure the reference tracking of the *SoC* (first part of the objective function). In the second part, the *SoC* variation is minimized to limit the charge and discharge of the battery. The third criterion minimizes the difference between the  $H_2$  system current seeing by the battery ( $H_2$  system current supplied to the battery) and the DC bus current seeing by batteries (current from batteries supplied to the DC bus). This is to bring the battery current to zero and to maintain the battery *SoC* when the *SoC* is close to its reference.

$$\begin{aligned} J(SoC_k) = \min_{i_{H2}^{bat}} & \sum_{k=0}^{N_{H2}} (SoC_k - SoC_k^*) Q_{r_{H2}} (SoC_k - SoC_k^*)^T \\ & + (SoC_k - SoC_{k-1}) Q_{y_{H2}} (SoC_k - SoC_{k-1})^T \\ & + \left( i_{H2,k}^{bat} - i_{l_{pv},k}^{bat} \right) \left( i_{H2,k}^{bat} - i_{l_{pv},k}^{bat} \right)^T \end{aligned} \quad (36)$$

- Parameters setting

The FC and El current references depend on the  $H_2$  system current reference value  $i_{H2}^{bat}$ . If  $i_{H2}^{bat} \geq 0$ , it is assigned to the FC,  $v_{H2} = v_{fc}$  and  $\Delta i_{H2,max} = \Delta i_{fc,max}$ . Vice versa if  $i_{H2}^{bat} \leq 0$  it is assigned to the El,  $v_{H2} = v_{el}$  and  $\Delta i_{H2,max} = \Delta i_{el,max}$ . The condition statement is summarized as shown on the Algorithm 1.

---

**Algorithm 1:** Fuel cell and electrolyzer current references

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```

1 if  $i_{H2}^{bat} \geq 0$  then
2    $v_{H2,k} = v_{fc,k}$ 
3    $\Delta i_{H2,max} = \Delta i_{fc,max}$ 
4 else
5    $v_{H2,k} = v_{el,k}$ 
6    $\Delta i_{H2,max} = \Delta i_{el,max}$ 

```

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The maximum and minimum  $H_2$  system currents are set to  $i_{H2,min} = i_{el,min}$ ,  $i_{H2,max} = i_{fc,max}$ .

### 2.2.3. Sub-Optimal Kalman Filter

Due to the current sensor cost and the distance between the load and the generator in a micro-grid, the load current is considered to be an unmeasured variable to reduce the number of sensors and improve the reliability. Here, it is estimated with a sub-optimal Kalman filter based on the measurement of the DC bus voltage.

$$v_{b,k+1} = v_{b,k} + \frac{T_{sbat}}{C_b} \left[ i_{bat,k}^b \right] - \frac{T_{sbat}}{C_b} \left[ i_{l,k} - i_{pv,k}^b - i_{H2,k}^b(k) \right] \quad (37)$$

where  $i_l - i_{pv}^b - i_{H2}^b$  is considered to be an unknown input called *fv*, which is assumed to be nearly constant during one sampling period, i.e.,  $\dot{fv} = 0$ . It follows that *fv* is estimated based on this linear state space model:

$$\underbrace{\begin{bmatrix} v_{b,k+1} \\ f v_{k+1} \end{bmatrix}}_{x_{obs,k+1}} = \underbrace{\begin{bmatrix} 1 & -\frac{T_{sbat}}{C_b} \\ 0 & 1 \end{bmatrix}}_{A_{d_{obs}}} \underbrace{\begin{bmatrix} v_{b,k} \\ f v_k \end{bmatrix}}_{x_{obs,k}} + \underbrace{\begin{bmatrix} -\frac{T_{sbat}}{C_b} \\ 0 \end{bmatrix}}_{B_{d_{obs,k}}} \underbrace{i_{bat,k}^b}_{u_{obs,k}} \quad (38)$$

$i_{l_{pv}} = i_l - i_{pv}^b$  is determined with  $i_{H2}^b = \frac{v_{H2}}{v_b} i_{H2}$  and the estimated value  $f v$ .  $i_{l_{pv}}$  can be used in Equation (10) and  $i_{l_{pv}}^{bat} = \frac{v_b}{v_{bat}} i_{l_{pv}}$  in (22) and (36).

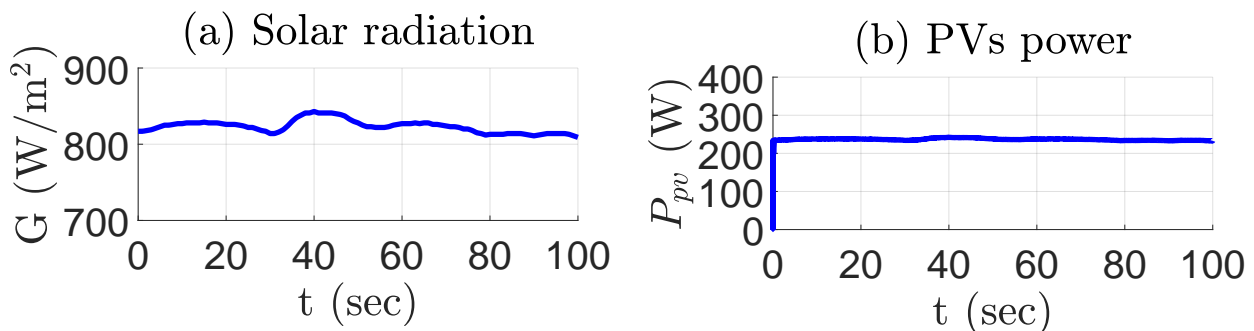
#### 2.2.4. Design Tools

The eMPCs are implemented with the Matlab (MathWorks, Natick, MA, USA toolbox Yalmip [33] and Multi-Parametric Toolbox 3 (MPT3) [34]. The optimization solver used is Gurobi Optimizer (Beaverton, USA) with Academic licence.

#### 2.3. Simulation Results

The system represented in Figure 1 is tested with the DeMPC to validate the control strategy before experimental tests. The system is simulated with a step time equal to 50  $\mu$ s. The SoC reference is equal to the initial SoC (50%) and the DC bus voltage reference is set to 70 V.

Figure 2 shows the PV system output power according to the solar radiation and the ambient air temperature (considered to be constant at 30 °C). The solar radiation data are based on a measurement campaign with a sampling step of 1s carried out on the UFR science building at the Moufia site of the University of La Réunion. Result of the PV output power simulation is consistent with the expected results.

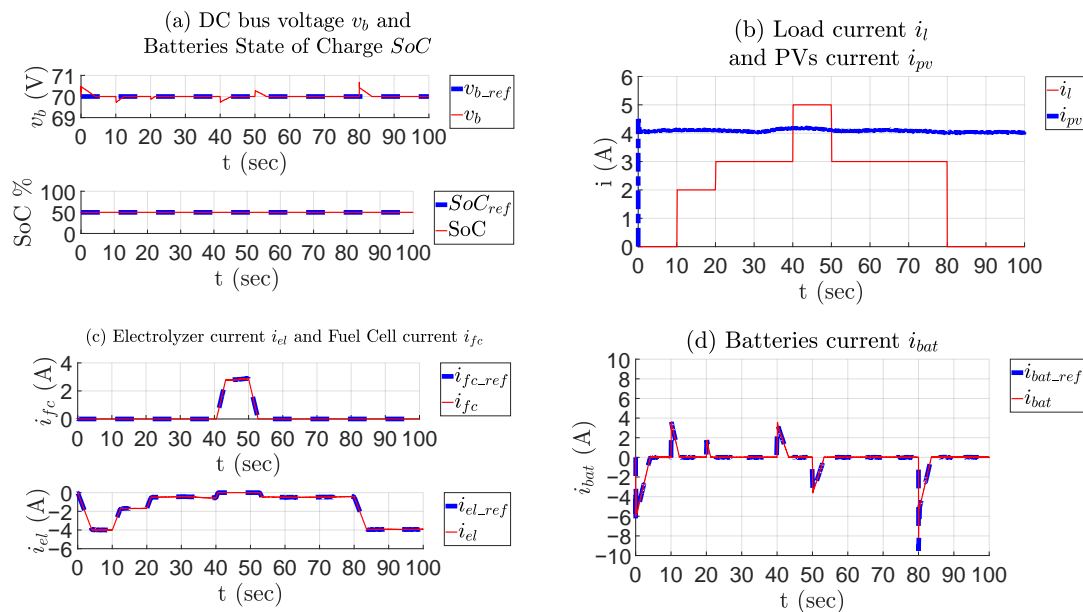


**Figure 2.** Simulation: (a) Solar radiation, (b) PV power.

Figure 3 represents the simulation results of the DeMPC and shows that the DC bus voltage reference tracking is correctly performed (between  $-0.4\%$  and  $+0.95\%$  of the reference). Also, the FC and El currents maximum ramp are respected, and batteries react to the high variation of the load or the PV production. The SoC variation is insignificant due to the short simulation time. Between 0 to 40 s and 50 to 100 s, the PV power is higher than the load power. In this duration, the El is running (sub-zero current). At 0, 50 and 80 s respectively, the DC bus voltage increases due to the surplus production. The battery absorbs excess production while the El current increases progressively in respect to the ramp rate current to bring back the battery current to zero. At 10, 20 and 40 s respectively, the battery provides the load when the load power increase and the El current decrease progressively in respect with the ramp rate current. This is to bring back the battery current to zero. In the opposite case, between 40 and 50 s, the PV power is lower than the load power and at 40 s the DC bus voltage decrease due to the production deficit. At 40 s, the battery provides the load while the FC current increases progressively in respect to the ramp rate current to bring back the battery current to 0. At 50 s, the battery absorbs the FC current to let it decrease progressively and bring back the battery current to 0. At 80 s, a

higher current peak can be observed. This peak is due to the higher difference between the PV current and the load current and due to the step time choose for the battery eMPC [32]. Simulation shows that the proposed strategy gives excellent results in terms of DC bus voltage regulation while taking into account elements operation constraints.

Compared to the simulation results in [13] with a mixed droop control strategy, the error on the DC bus voltage induced by a load current variation is lower and the FC, battery, and electrolyzer current variation is more linear.

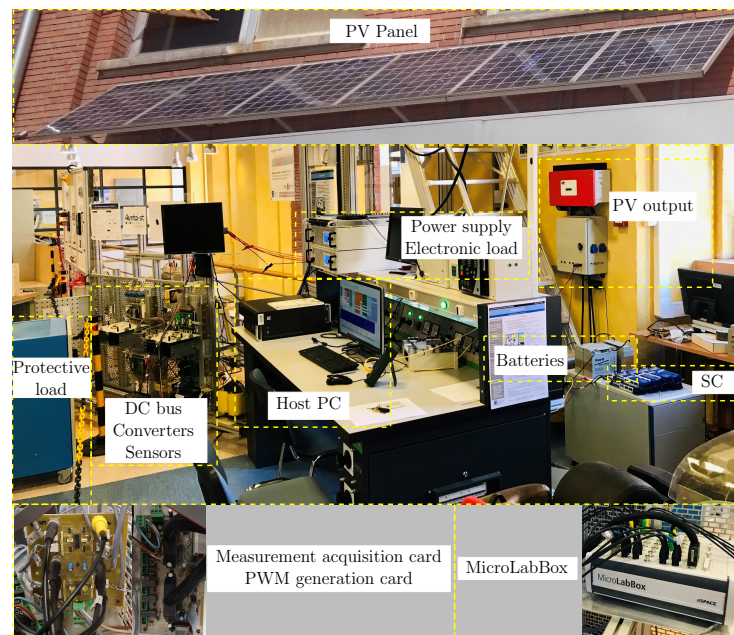


**Figure 3.** Simulation:  $SoC^* = SoC_{init}$ , (a) DC bus voltage and batteries  $SoC$ , (b) load and photovoltaics currents, (c) fuel cell and electrolyzer current and (d) batteries energy storage system current.

### 3. Experimentation

#### 3.1. Procedure

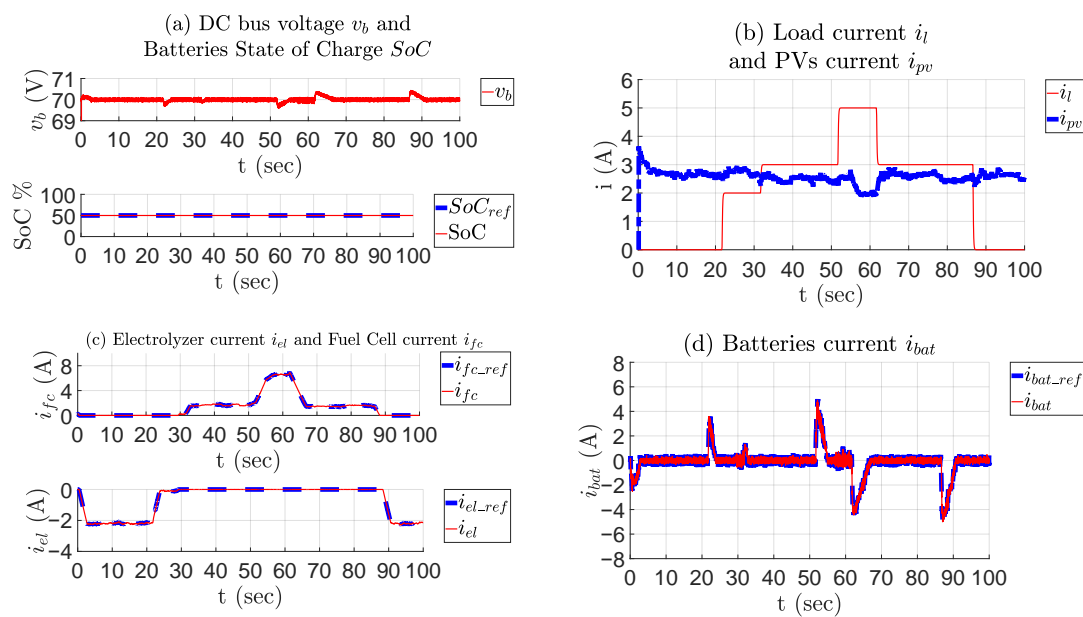
Experimental tests were carried out on a Power Hardware-In-the-Loop (PHIL) test bench shown in Figure 4. The parameters of the test bench system are available in Appendix A. It is composed of three 12 V AGM batteries and two 165 W PV modules with their converters. As a first step, to reduce the test bench cost and complexity, the fuel cell power output (FC and converter) is emulated with a programmable supply (reference: EA-PSI 9080-340). For the same reason, the power load requirement and the electrolyzer power load (EI and converter) are emulated with an electronic load (reference: EA-ELR 9080-510). The controller runs on a dSPACE DS1202 board and the measurements are made with a DS1302 board.



**Figure 4.** Power-Hardware-In-the-Loop test bench.

### 3.2. Results

At the experimentation beginning, the load current is null, and the PV is connected to the system. The load cycle begins when all currents are stabilized. As for the simulation, the DC bus voltage reference tracking is correctly performed (between  $-2.43\%$  and  $+0.62\%$  of the DC bus voltage reference, Cf. Figure 5). The experimental results show the control strategy ability to control a real system. The results are consistent with those expected after the simulation. The maximum FC and EI currents ramp rate are respected. From 0 to approximately 30 s and approximately 85 to 100 s, the PV power is higher than the load power. In this duration, the EI is running (sub-zero current). At 0 and between 80 and 90 s, the DC bus voltage increased due to the surplus production. The battery absorbs excess production while the EI current increases progressively in respect to the ramp rate current to bring back the battery current to zero. Between 20 and 30 s, the battery provides the load when the load power increase and the EI current decrease progressively in respect with the ramp rate current. This is to bring back the battery current to zero. In the opposite case, between 30 and 90 s, the PV power is lower than the load power and at 40s the DC bus voltage and decrease due to the production deficit. At 30 s and 50 s, the battery provides the load while the FC current increase progressively in respect to the ramp rate current to bring back the battery current to 0. At 60 s and between 80 and 90 s, the battery absorbs the FC current to let it decrease progressively. The higher oscillation that can be observed in the DC bus voltage and the battery current compared to the simulation is due to the order reduction made for the system equations.



**Figure 5.** Experimentation:  $SoC^* = SoC_{init}$ , (a) DC bus voltage and batteries SoC, (b) load and photovoltaics current, (c) fuel cell and electrolyzer current and (d) batteries energy storage system current.

#### 4. Discussion

Compared to the initial work in [20], PV panels and an electrolyzer have been added to the system to complete the work and test the strategy for a micro-grid configuration. The experimental results show the ability of the proposed strategy to manage an H2 system with hydrogen storage, hydrogen production from the electrolyzer and electricity production from the FC. Moreover, the fluctuations of the battery current are lower than those of the SC current in the initial work. Finally, compared to other control strategies in the literature, the use of eMPC does not result in any slowdown near the reference, such as regular controllers where the dynamic is governed by first- or second-order differential equations [13,16,32]. With the presented strategy, the current evolution is more rectilinear when the PV production or load consumption changes. The constraints on the current ramp rates are strictly respected for the FC and the electrolyzer to improve the remaining useful time. This feature allows the chemical and physical characteristics of the FC and the electrolyzer to be taken into account to use them more effectively. Moreover, the eMPC can be reconfigured on-line to take into account a modification of the power source topology in case of failure [23] or units (FC or El) commitment in case of power partitioning. In addition to this, on-line tuning parameters are also possible based on FC/El state of health founded on diagnosis/prognosis indicators [24].

#### 5. Conclusions

In this paper, a Power Management System architecture for a micro-grid composed of proton exchange membrane fuel cell, alkaline electrolyzer, li-ion batteries and photovoltaic panels has been designed. Based on distributed explicit model predictive control, the Power Management System defines current reference of the fuel cell, the electrolyzer and the batteries while explicitly taking into account all component operating conditions and constraints. The effectiveness of the control strategy is validated in simulation and on a Power Hardware-In-the-Loop test bench. The average error in the DC bus voltage in steady state is about 0.02% in simulation and the experimental results confirm this performance. The ability of the explicit model predictive control to ensure voltage regulation in the presence of disturbances is therefore confirmed. The respect of the constraints of the components according to the operating conditions (ramp current, etc.) is also confirmed.

Many perspectives of this work are considered. We can cite the integration of an Energy Management System (EMS) combined with the proposed PMS developed in this work. The main objective of the EMS is the optimization of the engagement plan of the FC and the electrolyzer to take into account operating constraints and external conditions: e.g., minimum/maximum operating currents, minimum operating time, load and PV power forecasting.

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## Appendix A. System Parameters

Batteries			
Nominal capacity	18 Ah	Max charging current	15 A
Nominal voltage	36V	Max discharging current	15 A
PVs (1000 W/m <sup>2</sup> )			
Power at Pmax	165 W	T <sub>SMPPPT</sub>	1 ms
Voltage at Pmax	68.8 V	Current at Pmax	4.8 A
Electrolyzer		Fuel Cell	
H <sub>2</sub> max production	1 Nm <sup>3</sup> /h	Power at Pmax	1200 W
Nominal voltage	86 V	Nominal voltage	26 V
Nominal current	120 A	Nominal current	46 A
DC/DC converter			
L <sub>fc</sub>	100 mH	L <sub>el</sub>	100 mH
L <sub>bat</sub>	1 mH	L <sub>pv</sub>	1 mH
DC bus			
C <sub>b</sub>	19.8 μF	DC bus voltage reference	70 V
MPC			
T <sub>sbat</sub>	500 μs	Q <sub>rH2</sub>	15 × 10 <sup>5</sup>
T <sub>sH2</sub>	2 ms	Q <sub>yH2</sub>	5 × 10 <sup>4</sup>
N <sub>bat</sub>	5	Q <sub>rbat</sub>	50
N <sub>H2</sub>	3		

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